## SP5055

2.5 GHz BI-DIRECTIONAL I ${ }^{2} \mathrm{C}$ BUS CONTROLLED SYNTHESISER

The SP5055 is a single chip frequency synthesiser designed for TV tuning systems. Control data is entered in the standard ${ }^{12} \mathrm{C}$ BUS format. The device contains 4 addressable current limited outputs and 4 addressable BiDirectional open collector ports one of which is a 3 Bit ADC. The information on these ports can be read via the $I^{2} \mathrm{C}$ BUS. The device has one fixed $\mathrm{I}^{2} \mathrm{C}$ BUS address and 3 programmable addresses, programmed by applying a specific input voltage to one of the current limited outputs. This enables 2 or more synthesisers to be used in a system.

The device is available in two variants: the SP5055 in 18lead plastic DIL (DP18) and the SP5055S in 16-lead miniature plastic DIL (MP16). See Features below for functional differences between the devices.

## FEATURES

- Complete 2.5 GHz Single Chip System
- Programmable via $1^{2} \mathrm{C}$ BUS
- Low power consumption (5V 65mA)
- Low Radiation
- Phase Lock Detector

Varactor Drive Amp Disable

- 8 Controllable Outputs, 4 Bi-Directional (SP5055)

6 Controllable Outputs, 4 Bi-Directional (SP5055S)

- 5 Level ADC

Variable ${ }^{12} \mathrm{C}$ BUS Address For Multi Tuner Applications
Full ESD Protection *

* Normal ESD handling procedures should be observed.


## APPLICATIONS

Satellite TV
High IF Cable Tuning Systems


Fig. 1 Pin connections - top view

## ORDERING INFORMATION

SP5055 DP - (18 lead Plastic package)
SP5055S MP - (16 lead Miniature Plastic package)

## SP5055

## ELECTRICAL CHARACTERISTICS

Test conditions (unless otherwise stated)
$\mathrm{T}_{\text {amb }}=-10^{\circ} \mathrm{C}$ to $80^{\circ} \mathrm{C}, \mathrm{V}_{\mathrm{CC}}=+4.75 \mathrm{~V}$ TO 5.25 V
All pin connections refer to DP package.
These characteristics are guaranteed by either production test or design. They apply within the specified ambient temperature and supply voltage unless otherwise stated.

| Characteristic | Pin | Value |  |  | Units | Conditions |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | Min | Typ | Max |  |  |
| Supply current <br> Prescaler Input Voltage <br> Prescaler Input Voltage | $\begin{gathered} 14 \\ 15,16 \\ 15,16 \end{gathered}$ | $\begin{gathered} 50 \\ 100 \end{gathered}$ | 65 | $\begin{gathered} 80 \\ 300 \\ 300 \end{gathered}$ | mA <br> $m V_{\text {RMS }}$ <br> $m V_{\text {RMS }}$ | $v_{C C}=5 \mathrm{~V}$ <br> 500 MHz to 1.8 GHz <br> 2.5 GHz , see Fig. 5 |
| Prescaler Input Impedance Input Capacitance | 15,16 |  | $\begin{gathered} 50 \\ 2 \end{gathered}$ |  | $\begin{gathered} \Omega \\ \mathrm{pF} \end{gathered}$ |  |
| SDA,SCL Input High Voltage Input Low Voltage Input High Current Input Low Current Leakage Current | $\begin{aligned} & 4,5 \\ & 4,5 \\ & 4,5 \\ & 4,5 \\ & 4,5 \end{aligned}$ | $\begin{aligned} & 3 \\ & 0 \end{aligned}$ |  | $\begin{gathered} 5.5 \\ 1.5 \\ 10 \\ -10 \\ 10 \end{gathered}$ | V <br> V <br> $\mu \mathrm{A}$ <br> $\mu \mathrm{A}$ <br> $\mu \mathrm{A}$ | Input Voltage $=\mathrm{V}_{\mathrm{CC}}$ <br> Input Voltage $=0 \mathrm{~V}$ <br> When $V_{C C}=0 V$ |
| SDA Output Voltage | 4 |  |  | 0.4 | V | $\mathrm{I}_{\text {sink }}=3 \mathrm{~mA}$ |
| Charge Pump Current Low | 1 |  | $\pm 50$ |  | $\mu \mathrm{A}$ | Byte 4 Bit $2=0$, Pin $1=2 \mathrm{~V}$ |
| Charge Pump Current High | 1 |  | $\pm 170$ |  | $\mu \mathrm{A}$ | Byte 4 Bit $2=1$, Pin $1=2 \mathrm{~V}$ |
| Charge Pump Output Leakage Current | 1 |  |  | $\pm 5$ | nA | Byte 4 Bit $4=1, \operatorname{Pin} 1=2 \mathrm{~V}$ |
| Charge Pump Drive Output Current | 18 | 500 |  |  | $\mu \mathrm{A}$ | $\mathrm{V}_{\text {pin } 18}=0.7 \mathrm{~V}$ |
| Charge Pump Amplifier Gain |  |  | 6400 |  |  |  |
| Recommended Crystal Series Resistance |  | 10 |  | 200 | $\Omega$ |  |
| Crystal Oscillator Drive Level |  |  | 40 |  | mVp-p |  |
| Crystal Oscillator Source Impedance | 2 |  | -400 |  | $\Omega$ | Nominal Spread $\pm 15 \%$ |
| Output Ports |  |  |  |  |  |  |
| P0-P3 Sink Current* | 13-10 | 0.7 | 1 | 1.5 | mA | $\mathrm{V}_{\text {OUT }}=12 \mathrm{~V}$ |
| P0-P3 Leakage Current ${ }^{\text {* }}$ | 13-10 |  |  | 10 | $\mu \mathrm{A}$ | $\mathrm{V}_{\text {OUT }}=13.2 \mathrm{~V}$ |
| P4-P7 Sink Current | 9-6 | 10 |  |  | mA | $\mathrm{V}_{\text {OUT }}=0.7 \mathrm{~V}$ |
| P4-P7 Leakage Current | 9-6 |  |  | 10 | $\mu \mathrm{A}$ | $\mathrm{V}_{\text {OUT }}=13.2 \mathrm{~V}$ |
| Input Ports |  |  |  |  |  |  |
| P3 Input Current High | 10 |  |  | + 10 | $\mu \mathrm{A}$ | $\mathrm{V}_{\text {pin } 10}=13.2 \mathrm{~V}$ |
| P3 Input Current Low | 10 |  |  | -10 | $\mu \mathrm{A}$ | $\mathrm{V}_{\text {pin } 10}=0 \mathrm{~V}$ |
| P4,P5,P7 Input Voltage Low | 9,8,6 |  |  | 0.8 | V |  |
| P4,P5, P7 Input Voltage High | 9,8,6 | 2.7 |  |  | V |  |
| P6 Input Current High | 7 |  |  | + 10 | $\mu \mathrm{A}$ | See Table 3 for ADC Levels |
| P6 Input Current Low | 7 |  |  | -10 | $\mu \mathrm{A}$ |  |

* Ports P1-P2 not present on the SP5055S.


Fig. 2 Block diagram of SP5055. (Ports P1-P2 not available on SP5055S)

## FUNCTIONAL DESCRIPTION

The SP5055 is programmed from an $I^{2} \mathrm{C}$ BUS. Data and Clock are fed in on the SDA and SCL lines respectively as defined by the ${ }^{2} \mathrm{C}$ BUS format. The synthesiser can either accept new data (write mode) or send data (read mode). The Tables in Fig. 3 illustrate the format of the data. The device can be programmed to respond to several addresses, which enables the use of more than one synthesiser in an $I^{2} \mathrm{C}$ BUS system. Table 4 shows how the address is selected by applying a voltage to P3. The last bit of the address byte (R/W) sets the device into read mode if it is high and write mode if it is low. When the SP5055 receives a correct address byte it pulls the SDA line low during the acknowledge period and during following acknowledge periods after further data bytes are programmed. When the SP5055 is programmed into the read mode the controlling device accepting the data must pull down the SDA line during the following acknowledge period to read another status byte.

## WRITE MODE (FREQUENCY SYNTHESIS)

When the device is in the write mode bytes $2+3$ select the synthesised frequency while bytes $4+5$ select the output port states and charge pump information.

Once the correct address is received and acknowledged, the first bit of the next Byte determines whether that Byte is interpreted as byte 2 or 4, a logic 0 for frequency information and a logic 1 for charge pump and output port information. Additional data bytes can be entered without the need to re-address the device until an ${ }^{2} \mathrm{C}$ stop condition is recognised. This allows a smooth frequency sweep for fine tuning or AFC purposes.

If the transmission of data is stopped mid-byte (e.g. by another device on the bus) then the previously programmed byte is maintained.

Frequency data from bytes 2 and 3 is stored in a 15 -bit shift register and is used to control the division ratio of the 15 -bit programmable divider which is preceded by a divide-by-16 prescaler and amplifier to give excellent sensitivity at the local oscillator input, see Fig. 5. The input impedance is shown in Fig 7.

The programmed frequency can be calculated by multiplying the programmed division ratio by 16 times the comparison frequency $F_{\text {comp }}$.

When frequency data is entered, the phase comparator, via the charge pump and varicap drive amplifier, adjusts the local oscillator control voltage until the output of the programmable divider is frequency and phase locked to the comparison frequency.

The reference frequency may be generated by an external source capacitively coupled into pin 2 or provided by an on-board 4 MHz crystal controlled oscillator.

Note - the comparison frequency is 7.8125 KHz when a 4 MHz reference is used.

Bit 2 of byte 4 of the programming data (CP) controls the current in the charge pump circuit, a logic 1 for $\pm 170 \mu \mathrm{~A}$ and a logic 0 for $\pm 50 \mu \mathrm{~A}$ allowing compensation for the variable tuning slope of the tuner and also to enable fast channel changes over the full band. Bit 4 of byte 4 (TO) disables the charge pump if set to a logic 1 . Bit 8 of byte 4 (OS) switches the charge pump drive amplifier's output off when it is set to a logic 1 . Bit 3 of byte 4 (T1) selects a test mode where the phase comparator inputs are available on P6 and P7, a logic 1 connects $F_{\text {comp }}$ to P6 and $F_{\text {div }}$ to P7.

Byte 5 programs the output ports P0 to P7; a logic 0 for a high impedance output and a logic 1 for low impedance (on).

## READ MODE

When the device is in the read mode the status data read from the device on the SDA line takes the form shown in Table 2

Bit 1 (POR) is the power-on reset indicator and is set to a logic 1 if the power supply to the device has dropped below 3 V and the programmed information lost (e.g., when the device is initially turned on). The POR is set to 0 when the read sequence is terminated by a stop command. The outputs are all set to high impedance when the device is initially powered up. Bit 2 (FL) indicates whether the device is phase locked, a logic 1 is present if the device is locked and a logic 0 if the device is unlocked.

Bits 3,4 and $5(12, I 1, I 0)$ show the status of the I/O Ports P7, P5 and P4 respectively. A logic 0 indicates a low level and a logic 1 a high level. If the ports are to be used as inputs they should be programmed to a high impedance state (logic 1). These inputs will then respond to data complying with TTL type voltage levels. Bits 6,7 and 8 (A2, A1, A0) combine to give the output of the 5 level ADC.

The 5 level ADC can be used to feed AFC information to the microprocessor from the IF section of the receiver, as illustrated in the typical application circuit.

## APPLICATION

A typical Application is shown in Fig. 4. All input/ output interface circuits are shown in Fig. 6.

| MSB |  |  |  |
| :--- | :---: | :---: | :---: |
| ADDRESS 1 1 0 0 0 MA1 MA0 0 A BYTE 1 <br> PROGRAMMABLE <br> DIVIDER 0 $2^{14}$ $2^{13}$ $2^{12}$ $2^{11}$ $2^{10}$ $2^{9}$ $2^{8}$ A BYTE 2 <br> PROGRAMMABLE <br> DIVIDER $2^{7}$ $2^{6}$ $2^{5}$ $2^{4}$ $2^{3}$ $2^{2}$ $2^{1}$ $2^{0}$ A BYTE 3 <br> CHARGE PUMP <br> AND TEST BITS 1 CP T1 T0 1 1 1 OS A BYTE 4 <br> IO PORT CONTROL <br> BITS P7 P6 P5 P4 P3 P2* P1 $^{*}$ P0 A BYTE 5 |  |  |  |

Table 1 Write data format (MSB is transmitted first)

| ADDRESS | 1 | 1 | 0 | 0 | 0 | MA1 | MAO | 1 | A | BYTE 1 |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| STATUS BYTE | POR | FL | 12 | 11 | 10 | A2 | A1 | A0 | A | BYTE 2 |

Table 2 Read data format

A : Acknowledge Bit
MA1, MA0 : Variable address bits (see Table 4)
CP : Charge Pump current select
T1: Test mode selection
T0: Charge pump disable
OS: Varactor drive Output disable Switch
P7, P6, P5, P4, P3, P2 ${ }^{\star}, \mathrm{P}^{\star}, \mathrm{P} 0$ : control output states
POR : Power On Reset indicator
FL : Phase Lock detect Flag
12, I1, 10 : Digital information from Ports P7, P5 and P4, respectively
A2, A1, A0:5 Level ADC data from P6 (see Table 3)

| $A 2$ | $A 1$ | $A 0$ | Voltage input to P 6 |
| :--- | :--- | :--- | :--- |
| 1 | 0 | 0 | $0.6 \mathrm{~V}_{C C}$ to 13.2 V |
| 0 | 1 | 1 | $0.45 \mathrm{~V}_{\mathrm{CC}}$ to $0.6 \mathrm{~V}_{\mathrm{CC}}$ |
| 0 | 1 | 0 | $0.3 \mathrm{~V}_{\mathrm{CC}}$ to $0.45 \mathrm{~V}_{\mathrm{CC}}$ |
| 0 | 0 | 1 | $0.15 \mathrm{~V}_{\mathrm{CC}}$ to $0.3 \mathrm{~V}_{\mathrm{CC}}$ |
| 0 | 0 | 0 | 0 to $0.15 \mathrm{~V}_{\mathrm{CC}}$ |


| MA1 | MA0 | Voltage input to P 3 |
| :---: | :---: | :---: |
| 0 | 0 | $0-0.2 \mathrm{~V}_{\mathrm{CC}}$ |
| 0 | 1 | ALWAYS VALID |
| 1 | 0 | $0.3-0.7 \mathrm{~V}_{\mathrm{CC}}$ |
| 1 | 1 | $0.8 \mathrm{~V}_{\mathrm{CC}}-13.2 \mathrm{~V}$ |

Table 4 Address selection

Table 3 ADC levels

NOTE: * Don't care condition on SP5055S
Fig. 3 Data Formats


Fig. 4 Typical SP5055 application


Fig. 5 Typical input sensitivity


RF input


Ports P7-P4


Reference oscillator


Loop amplifier



Fig. 7 Typical input impedance

## ABSOLUTE MAXIMUM RATINGS

All voltages are referred to $\mathrm{V}_{\mathrm{EE}}$ and pin 3 at 0 V .

| Parameter | $\begin{gathered} \text { Pin } \\ \text { SP5055 } \end{gathered}$ | $\begin{gathered} \text { Pin } \\ \text { SP5055S } \end{gathered}$ | Value |  | Units | Conditions |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | Min | Max |  |  |
| Supply voltage | 14 | 12 | -0.3 | 7 | V |  |
| RF input voltage | 15, 16 | 13, 14 |  | 2.5 | Vp-p |  |
| Port voltage | 6-13 | 6-11 | -0.3 | 14 | v | Port in off state |
|  | $6-9$ $10-13$ | $6-9$ 10,11 | -0.3 -0.3 | 6 14 | V | Port in on state Port in on state |
| Total port output current | 6-13 | 6-11 |  | 50 | mA |  |
| RF input DC offset | 15, 16 | 13, 14 | -0.3 | $\mathrm{V}_{\mathrm{CC}}+0.3$ | V |  |
| Charge pump DC offset | 1 | 1 | -0.3 | $V_{C C}+0.3$ | V |  |
| Drive DC offset | 18 | 16 | -0.3 | $V_{C C}+0.3$ | v |  |
| Crystal oscillator DC offset | 2 | 2 | -0.3 | $V_{C C}+0.3$ | V |  |
| SDA, SCL input voltage | 4, 5 | 4, 5 | $\begin{aligned} & -0.3 \\ & -0.3 \end{aligned}$ | $\underset{5.5}{\mathrm{v}_{\mathrm{CC}}+0.3}$ | $\begin{aligned} & \text { V } \\ & \text { V } \end{aligned}$ | With $V_{C C}$ applied $V_{C C}$ not applied |
| Storage temperature |  |  | -55 | + 125 | ${ }^{\circ} \mathrm{C}$ |  |
| Junction temperature |  |  |  | + 150 | ${ }^{\circ} \mathrm{C}$ |  |
| DP18 thermal resistance, chip-to-ambient DP18 thermal resistance, chip-to-case |  |  |  | 78 24 | $\begin{aligned} & { }^{\circ} \mathrm{C} / \mathrm{W} \\ & { }^{\circ} \mathrm{C} / \mathrm{W} \end{aligned}$ |  |
| MP16 thermal resistance, chip-to-ambient MP16 thermal resistance, chip-to-case |  |  |  | $\begin{gathered} 111 \\ 41 \end{gathered}$ | $\begin{aligned} & { }^{\circ} \mathrm{C} / \mathrm{W} \\ & { }^{\circ} \mathrm{C} / \mathrm{W} \end{aligned}$ |  |
| Power consumption at 5.5 V |  |  |  | 440 | mW | All ports off |

